

The modulus of elasticity  $E^E$  describes the mechanical elasticity of the piezo-material for full decoupling, as it is achievable for any value of  $\alpha$  if the electrodes on the piezo sheet are shorted. For electrical measurements, on the other hand, the decoupling occurs for every value of  $\alpha$  if any movement is prevented. As a thought experiment: for  $v = 0$ , the secondary current of the transducer is zero, and therefore only the capacitance of the crystal  $C_K$  remains. In literature, this special case is designated with a superscript S (clamped mechanical relative deformation S). Typical values for the relative dielectric constant are  $\epsilon_r^S = 1000 \dots 4000$ . Given the above, the capacitance of the crystal  $C_K$  can be calculated:

$$C_K = \epsilon_0 \cdot \epsilon_r^S \cdot S / h; \quad \epsilon_0 = 8,85 \cdot 10^{-12} \text{ As/Vm} \quad \text{Crystal-capacitance}$$

For regular (unclamped) operation, we measure – on the electrical side – two capacitances connected in parallel: the *crystal-capacitance*  $C_K$ , and the *spring capacitance*  $C_s$  caused by the mechanical side:

$$C_s = \alpha^2 / s_K \quad \text{Spring-capacitance}$$

**In summary:** For electrical no-load, i.e. the open-circuit situation with only  $C_K$  having an effect on the electrical side, we measure two stiffnesses on the mechanical side:  $s = s_K + s_C$ . For mechanical no-load with only  $s_K$  having an effect on the mechanical side, we measure two capacitances on the electrical side:  $C = C_K + C_s$ . The reactive load that is transformed by the transducer to the respective other side is:  $C_s / C_K = s_C / s_K \approx 50 \dots 100\%$ . Given high-grade electro-mechanical or mechano-electrical linkage\*, we may assume  $C_s \approx C_K$ ; depending on the piezo material, lower values are possible, as well.

## 6.2 Electrical loading

An open-circuit connection at the transducer represents an **idealization** that does not really exist in this form. For a piezo-electric guitar pickup, the electrical side is loaded via the cable (acting as a capacitance) and the input impedance of the amplifier, while on the mechanical side, the bridge and the strings need to be considered. Let us assume as a first approach an imprinted force  $\underline{F}$  to be the mechanical source. In order to calculate the output voltage, the simplest approach is to transform both this force and the crystal stiffness onto the electrical side [3]:

$$\underline{U} = \underline{F} / \alpha \quad C_s = \alpha^2 / s_K \quad \text{Transformed quantities}$$

This transformation yields a purely electrical network that may be investigated using the known approaches for network analysis. Of particular importance is the effect of the electrical load impedance on the **transmission function**  $\underline{H}$ . The input impedance of a guitar amplifier typically amounts to 1 M $\Omega$ ; relative to this, a line input can be of much lower impedance (e.g. 50 k $\Omega$ ). The capacitive internal impedance of the pickup forms – on cooperation with the input impedance of the amplifier – a first-order high-pass (**Fig. 6.2**).

\* Linkage-factor:  $k^2 = W_{\text{mech}} / W_{\Sigma} = C_s / (C_s + C_K) = 0,3 \dots 0,5$ .

Seen from the side of the load resistance, the two capacitances are connected in parallel, and they therefore are added up to calculate the cutoff-frequency of the high-pass (3-dB-frequency):  $C = C_K + C_s$ .

$$f_g = \frac{1}{2\pi R(C_K + C_s)} = \frac{1}{2\pi RC} \quad \text{Cutoff-frequency of the high-pass}$$

With  $C = 1,5 \text{ nF}$  and  $R = 1 \text{ M}\Omega$ , we obtain  $f_g = 106 \text{ Hz}$ , which is a value matching the frequency range of the guitar. For  $R = 50 \text{ k}\Omega$  (line input), the cutoff frequency would rise to  $2,1 \text{ kHz}$ , corresponding to a complete loss of the lows. Even less suitable would be a microphone input: its input impedance usually amounts to only about  $2 \text{ k}\Omega$ .

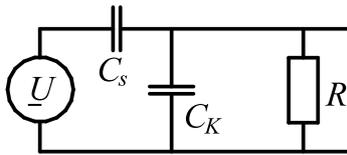


Fig. 6.2: Equivalent circuit of the transducer

For an **active** piezo pickup, the input impedance of the guitar amplifier is immaterial – a battery-operated preamplifier with low output impedance is integrated into the guitar, and makes for a problem-free connection to line inputs. A **passive** piezo system, however, does not include a preamplifier, and the pickup-signal needs to be fed to the guitar amplifier using a shielded cable. High-quality cables represent merely a capacitive load (Chapter 9.4), and therefore increase  $C_K$ . On the one hand, this has the effect of a broadband signal attenuation (capacitive divider), on the other hand it decreases the cutoff frequency of the high-pass. The ubiquitous assumption that a **long cable** would attenuate in particular the treble range does not hold for the piezo pickup – the internal impedance of the latter does not have a resistive character but a capacitive one.

### 6.3 The piezo-transducer as a sensor

In its operation as a sensor, the guitar pickup converts mechanical input signals (string vibrations) into electrical output signals. This represents the normal case; an operation as actor – which is also possible – is of interest only in the context of metrology (Chapter 6.5). According to the idealization used so far, the piezo transducer works as force-to-voltage converter. It captures in particular the AC-component of the bearing force the strings cause, and generates a correspondingly proportional voltage. The bearing force does not act directly onto the piezo crystal, though – a **pickup housing**, with its masses and springs, represents a mechanical filter. The effects of this filter will be investigated from a metrology-point-of-view in the following, using an Ovation pickup (Fig. 6.3) as example.

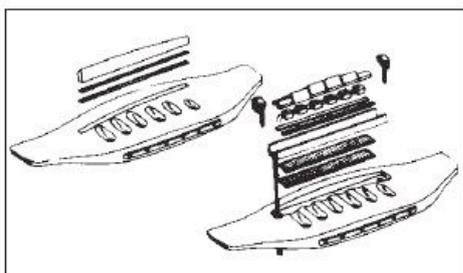


Fig. 6.3: Two different piezo pickups (Ovation).